

Radar Tutorial Solution - Flt Lt Woodward

These notes aim to document my thought process for the design of the surveillance radar tutorial.

Wavelength for a 5GHz signal.

```
In[65]:=  $\lambda = \frac{3 \times 10^8}{5. \times 10^9}$ 
```

```
Out[65]= 0.06
```

Height required at the radar antenna for a 45km range with a tgt at 75m.

```
In[26]:=  $\left( \frac{45}{4.1} - \sqrt{75} \right)^2$ 
```

```
Out[26]= 5.36087
```

Even though the radar being 5.4m up will give a line of sight, the ground will impinge on the Fresnel zone, so we need to do something about this later.

15 degree beam width converted to radians.

```
In[30]:=  $15. / 360 \times 2 \text{ Pi}$ 
```

```
Out[30]= 0.261799
```


Maximum vertical dimension of the antenna to achieve a 15 degree beam width in elevation. $\frac{\lambda}{\theta}$

```
In[34]:=  $0.06 / 0.262$ 
```

```
Out[34]= 0.229008
```

Calculate the maximum height covered by a 15 degree beam at max range. $\text{Tan}\theta = \text{Opp} / \text{Adj}$.

```
In[36]:=  $45. \times 10^3 \text{ Tan}[15 \text{ Degree}]$ 
```

```
In[37]:=  UnitConvert[12060 m, "Feet"]
```

```
Out[37]=  $\frac{5025000}{127} \text{ ft}$ 
```

```
In[38]:=  $\text{N}\left[\frac{5025000}{127} \text{ ft}, 6\right]$ 
```

```
Out[38]= 39566.9 ft
```

We assumed a horizontal dimension to the antenna of 3m, so we no compute the azimuth beam width.

```
In[39]:=  $0.06 / 3$ 
```

```
Out[39]= 0.02
```

$$\text{In[40]:= } \frac{0.02}{2 \text{ Pi}} 360$$

$$\text{Out[40]:= } 1.14592$$

Given both of the beam widths, we can calculate the antenna gain, assuming a rectangle.

$$\text{In[41]:= } \frac{42000.}{15 \times 1.15}$$

$$\text{Out[41]:= } 2434.78$$

and convert to dBs. (for a sanity check)

$$\text{In[44]:= } \mathbf{N[10 \text{ Log}[10, 2434]]}$$

$$\text{Out[44]:= } 33.8632$$

Next, the probability of detection...

$$0.98 \text{ (from the spec)} = \binom{4}{3} P_D^3 (1 - P_D) + \binom{4}{4} P_D^4 (1 - P_D)^0$$

$$0.98 = 4 \cdot P_D^3 (1 - P_D) + 1 \cdot P_D^4$$

$$\text{In[46]:= } \mathbf{Binomial[4, 3]}$$

$$\text{Out[46]:= } 4$$

$$\text{In[48]:= } \mathbf{Solve[4 x^3 - 3 x^4 == 0.98, x]}$$

$$\text{Out[48]:= } \{ \{x \rightarrow -0.331096 - 0.468634 i\}, \{x \rightarrow -0.331096 + 0.468634 i\}, \{x \rightarrow 0.93986\}, \{x \rightarrow 1.05566\} \}$$

The only relevant solution is 0.9398, or 0.94.

Given this probability of detection for one scan, and an assumed false alarm rate of 10^{-6} , we look up the required SNR from the graph. 13.5 dB. Converted to linear numbers.

$$\text{In[57]:= } \mathbf{10^{\frac{13.5}{10}}}$$

$$\text{Out[57]:= } 22.3872$$

Next, Band width, $c / 2 \cdot \Delta R$

$$\text{In[49]:= } \frac{3 \times 10^8}{2 \times 30.}$$

$$\text{Out[49]:= } 5. \times 10^6$$

This goes together with an uncompressed pulse width.

$$\text{In[70]:= } \frac{1.}{5 \times 10^6}$$

$$\text{Out[70]:= } 2. \times 10^{-7}$$

or 0.2 μSec .

Un-ambiguous range PRF

$$\text{In[50]:= } \frac{3 \times 10^8}{2. \times 45 \times 10^3}$$

$$\text{Out[50]:= } 3333.33$$

We assumed a RPM of 15, i.e. a 4 sec scan. This seems reasonable, given the furthest a target can progress is $4 \times 200 \text{ ms}^{-1}$ or 800m and

In[91]:= **45 000 / 800 .**

Out[91]= 56.25

this gives 56 seconds to detect it.

Number of pulses in each beam dwell, PRF x beam width / 6. RPM.

In[80]:=
$$\frac{3333 \times 1.15}{6 \times 15}$$

Out[80]= 42.5883

We're now in a position to make our first estimate at the peak power output required.

$$P_t = \frac{R^4 (4\pi)^3 K. T. B. SNR}{G^2 \lambda^2 \sigma n E_i}$$

Additional assumptions:

Temp is 1000k to allow for the receiver noise and noise figure.

E_i is 0.8 (data sheet suggests between 0.7 and 0.9 for a coherent receiver).

In[62]:=
$$\frac{45\,000^4 (4\pi)^3 1.38 \times 10^{-23} 1000 \times 5 \times 10^6 22.4}{2434^2 0.06^2 10 \times 42 \times 0.8}$$

Out[62]= 1755.07

This regime has a duty cycle of Pulse width x PRF.

In[72]:= **$2. \times 10^{-7} 3333$**

Out[72]= 0.0006666

This duty cycle is way below what is achievable. i.e. there is much scope for higher PRF and / or longer pulses. However, a longer pulse will require pulse compression to maintain range resolution and a higher PRF will require PRF agility (probably stagger) to maintain the un-ambiguous range. Increasing the pulse width has the added advantage of reducing the peak power by spreading it over a longer time period to maintain the same energy.

This does not take into account rain, or other weather phenomena, or any losses. Put simply, peak power output could easily be double or more, but we have no means of making an assessment with the given data. Clutter, however, will be a problem, so we need to make assessment of the possibility of a MTI mode. We need to choose a PRF so that the blind speed of the radar will be greater than 200 ms^{-1} .

Blind velocities $V_{\text{blind}} = \frac{n \cdot \lambda \text{PRF}}{2}$ where $n \in \mathbb{Z}^+$. We're concerned with the first, i.e. $n = 1$.

In[68]:=
$$\frac{0.06 \times 3333}{2}$$

Out[68]= 99.99

so we would need to double our PRF for the max speed to correspond to the MTI first blind speed. i.e. we would need a PRF of 10000Hz to be really clear. The max un-ambiguous range for a PRF of 10KHz is

In[73]:=
$$\frac{3. \times 10^8}{2. \times 10\,000}$$

Out[73]= 15 000.

i.e. 15Km, well below what is needed. If I use a 2 position stagger with a PRF of 10KHz and a PRF with max unambiguous range of 9km, the combination will have a LCM of 45km as required.

$$\text{In[74]:= } \frac{3 \times 10^8}{2. \times 9 \times 10^3}$$

Out[74]= 16 666.7

If I also alternate them in a pair wise fashion I end up with a 2 element, 2 position stagger. Alternatively, I could use a 2 element 5 position stagger, to increase the pulse density. ie, PRF 1 is 10KHz, PRF 2 is 16.667KHz. and I fire 1,2,1,1,2. This sequence has the total duration of $2/10000 + 3/16667$

$$\text{In[77]:= } 2. / 10\ 000 + 3 / 16\ 667$$

Out[77]= 0.000379996

and the average PRF is

$$\text{In[79]:= } \text{Mean}\{10\ 000., 16\ 667, 10\ 000, 16\ 667, 16\ 667\}$$

Out[79]= 14 000.2

and this pattern can fit into the beam dwell time ... times.

$$\text{In[81]:= } \frac{14\ 000 \times 1.15}{6 \times 15}$$

Out[81]= 178.889

this has the duty cycle of

$$\text{In[84]:= } 2. \times 10^{-7} \times 14\ 000$$

Out[84]= 0.0028

i.e. approx 0.3%. This can be increased further, so we now consider pulse compression, to maintain the range resolution, but increase the pulse width. If we aim for a 10% duty cycle, I can increase my pulse duration by $10/0.28$

$$\text{In[85]:= } 10 / 0.28$$

Out[85]= 35.7143

$$\text{In[87]:= } 2 \times 10^{-7} * 35.$$

Out[87]= $7. \times 10^{-6}$

The pulse is now $7\mu\text{s}$, but the band width is still the same, so I compute γ , the rate of linear chirp required. So if we have a compression ratio of 35, the change in frequency required over the new $7\mu\text{s}$ pulse is

$$\text{In[89]:= } \frac{35}{7. \times 10^{-6}}$$

Out[89]= $5. \times 10^6$

i.e. a linear chirp of 5MHz. Because of the increase in noise in the receiver due to the longer pulse, this has no effect on the detection. However, we can re-compute the range equation for the revised figures.

$$\text{In[90]:= } \frac{45\ 000^4 (4 \text{ Pi})^3 1.38 \times 10^{-23} 1000 \times 5 \times 10^6 22.4}{2434^2 0.06^2 10 \times 179 \times 0.8}$$

Out[90]= 411.804

So we have a peak power requirement of 411 watts (same caveates as before - probably more like 1kW), and with the new PRF pattern and Chirp, a 10% duty cycle.